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jc839 U.S. PTO
09/27/00

PATENT APPLICATION
IN THE UNITED STATES PATENT AND TRADEMARK OFFICE
UTILITY PATENT APPLICATION TRANSMITTAL LETTER

jc784 U.S. PTO
09/27/00
09/670870

Atty./Agent Docket No.: CM00914S

Mailing Date: 9-27-00

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Assistant Commissioner for Patents
Box Patent Application
Washington, D.C. 20231

Dear Sir:

Transmitted herewith for filing under 37 CFR 1.53 (b) is a Nonprovisional Utility Patent Application:

X New Application; or

 Continuation; or Divisional, or Continuation-in-Part (CIP) Application of prior US application No. , filed on , having US Examiner , in Group Art Unit : of

Inventor(s): Shraga, Boaz Or et al.

For (Title): ECHO SUPPRESSION AND ECHO CANCELLATION

This transmittal letter has 2 total pages.

Enclosed are:

X 3 sheets of formal drawings, along with 39 pages of specification, claims, and abstract.

X Oath or Declaration Combined with Power of Attorney (3 pages)

X Newly Executed (original or copy)

 Copy from a prior application (if this is a Continuation/Division with no new matter)

 Statement deleting named inventor(s) in prior application if this is a

 Continuation/Division (See 37 CFR 1.63(d)(2) and 1.33(b).)

 Consider as the above Statement, Please delete as inventors for this application the following inventors named in the prior application:

X Foreign priority to Great Britain Patent application having serial number 9926731.2 and a filing date of November 11, 1999 is hereby claimed under 35 USC 119.

X A copy of the priority document is included herewith.

X An Assignment Transmittal Letter and Assignment of the invention to MOTOROLA, INC.

 An Information Disclosure Statement (IDS), with PTO-1449, and citation copies.

 Petition For Extension of Time for parent application of the present Continuation/Division/CIP application

X Print EFS Inventor Information Sheet(s).

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_____ Preliminary Amendment

_____ Please cancel filed claims _____

_____ Incorporation by Reference (for Continuation/Division application) The entire disclosure of the prior application, from which a copy of the oath or declaration is supplied, is considered as being part of the disclosure of the accompanying application and is hereby incorporated by reference therein.

_____ Since the present application is based on a prior US application, please amend the specification by adding the following sentence before the first sentence of the specification: "The present application is based on prior US application No. _____, filed on _____, which is hereby incorporated by reference, and priority thereto for common subject matter is hereby claimed."

X The filing fee is calculated as follows:

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MULTIPLE DEPENDENT CLAIMS			\$260	= \$ 0.00
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Parameter	Value	Unit
Initial concentration	1.0	g/L
Initial pH	7.0	
Temperature	25	°C
Time	0, 1, 2, 4, 8, 16, 32, 64, 128, 256, 512, 1024, 2048, 4096, 8192, 16384, 32768, 65536, 131072, 262144, 524288, 1048576, 2097152, 4194304, 8388608, 16777216, 33554432, 67108864, 134217728, 268435456, 536870912, 1073741824, 2147483648, 4294967296, 8589934592, 17179869184, 34359738368, 68719476736, 137438953472, 274877906944, 549755813888, 1099511627776, 2199023255552, 4398046511104, 8796093022208, 17592186044416, 35184372088832, 70368744177664, 140737488355328, 281474976710656, 562949953421312, 1125899906842624, 2251799813685248, 4503599627370496, 9007199254740992, 18014398509481984, 36028797018963968, 72057594037927936, 144115188075855872, 288230376151711744, 576460752303423488, 1152921504606846976, 2305843009213693952, 4611686018427387904, 9223372036854775808, 18446744073709551616, 36893488147419103232, 73786976294838206464, 147573952589676412928, 295147905179352825856, 590295810358705651712, 1180591620717411303424, 2361183241434822606848, 4722366482869645213696, 9444732965739290427392, 18889465931478580854784, 37778931862957161709568, 75557863725914323419136, 151115727451828646838272, 302231454903657293676544, 604462909807314587353088, 1208925819614629174706176, 2417851639229258349412352, 4835703278458516698824704, 9671406556917033397649408, 19342813113834066795298816, 38685626227668133590597632, 77371252455336267181195264, 154742504910672534362390528, 309485009821345068724781056, 618970019642690137449562112, 1237940039285380274899124224, 2475880078570760549798248448, 4951760157141521099596496896, 9903520314283042199192993792, 19807040628566084398385987584, 39614081257132168796771975168, 79228162514264337593543950336, 158456325028528675187087900672, 316912650057057350374175801344, 633825300114114700748351602688, 1267650600228229401496703205376, 2535301200456458802993406410752, 5070602400912917605986812821504, 10141204801825835211973625643008, 20282409603651670423947251286016, 40564819207303340847894502572032, 81129638414606681695789005144064, 162259276829213363391578010288128, 324518553658426726783156020576256, 649037107316853453566312041152512, 1298074214633706907132624082305024, 2596148429267413814265248164610048, 5192296858534827628530496329220096, 10384593717069655257060992658440192, 20769187434139310514121985316880384, 41538374868278621028243970633760768, 83076749736557242056487941267521536, 166153499473114484112975882535043072, 332306998946228968225951765070086144, 664613997892457936451903530140172288, 1329227995784915872903807060280344576, 2658455991569831745807614120560689152, 5316911983139663491615228241121378304, 10633823966279326983230456482242756608, 21267647932558653966460912964485513216, 42535295865117307932921825928971026432, 85070591730234615865843651857942052864, 170141183460469231731687303715884105728, 340282366920938463463374607431768211456, 680564733841876926926749214863536422912, 1361129467683753853853498429727072845824, 2722258935367507707706996859454145691648, 5444517870735015415413993718908291383296, 10889035741470030830827987437816582766592, 21778071482940061661655974875633165533184, 43556142965880123323311949751266331066368, 87112285931760246646623899502532662132736, 174224571863520493293247799005065324265472, 348449143727040986586495598010130648530944, 696898287454081973172991196020261297061888, 1393796574908163946345982392040522594123776, 2787593149816327892691964784081045188247552, 5575186299632655785383929568162090376495104, 11150372599265311570767859136324180752990208, 22300745198530623141535718272648361505980416, 44601490397061246283071436545296723011960832, 89202980794122492566142873090593446023921664, 178405961588244985132285746181186892047843328, 356811923176489970264571492362373784095686656, 713623846352979940529142984724747568191373312, 1427247692705959881058285969449495136382746624, 2854495385411919762116571938898990272765493248, 5708990770823839524233143877797980545530986496, 11417981541647679048466287755595961091061972992, 2283596308329	

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ECHO SUPPRESSION AND ECHO CANCELLATION

FIELD OF THE INVENTION

The present invention relates to methods and systems for
5 processing sound signals in general, and to methods and systems
for processing audio signals in the presence of echo conditions, in
particular.

BACKGROUND OF THE INVENTION

10 Echo cancellation and suppression systems and methods
therefor are known in the art. A speech transmission oriented
system, such as a speech communication system, includes a sound
production unit such as a speaker and a sound detection unit such
as a microphone. The speaker produces sounds received from a
15 remote source and the microphone detects voice sounds which are
provided by the user. It is noted that the microphone also detects
sounds from other sources such as noise and sounds, which are
produced by the speaker (echo), after traveling through an acoustic
path therebetween.

20 Methods and systems for "emphasizing" the voice portion
over the noise and echo are known in the art. The common

approaches use adaptive filtering to derive a replica of an echo-signal, which is further subtracted from the microphone signal. As a result, a level of the echo-signal is decreased significantly. Speech communication systems, especially hand-held devices, are limited in
5 their processing and storage resources. This fact forces developers to search for new approaches for robust and high quality echo cancellation and echo suppression methods, which can function within these limited conditions.

In the article by A. Hirano et al. "A noise-robust stochastic
10 gradient algorithm with an adaptive step-size suitable for mobile hands-free telephones", Proc. ICASST-95, v.5, pp. 1392-1395, 1995, an adaptive step-size algorithm is proposed. The algorithm controls the step size of the standard NLMS method, based on the reference input signal power and the noise power.

15 US Patent No 5,608,804 to Hirano, entitled "Method of and apparatus for identifying a system with adaptive filter" is directed to a method of and an apparatus for estimating characteristics of an unknown system, using an adaptive filter, in an echo canceller. A gradient step size, which is controlled dynamically, is a function of
20 the power of the reference input signal. The gradient step size monotonously increases if the power of the reference input signal is

smaller than a threshold and monotonously decreases if the power of the reference input signal is greater than the threshold.

US Patent No 5,546,459 to Sih, et al., entitled "Variable block size adaptation algorithm for noise-robust acoustic echo cancellation" is directed to an apparatus for acoustic echo
5 cancellation using an adaptive filter. The apparatus updates coefficients of the adaptive filter using a signal block of length L. The block size is adjusted in response to the instantaneous signal-to-noise ratio.

10

	1980	1981	1982	1983	1984	1985	1986	1987	1988	1989	1990	1991	1992	1993	1994	1995	1996	1997	1998	1999	2000	2001	2002	2003	2004	2005	2006	2007	2008	2009	2010	2011	2012	2013	2014	2015	2016	2017	2018	2019	2020	2021	2022	2023	2024	2025	2026	2027	2028	2029	2030	2031	2032	2033	2034	2035	2036	2037	2038	2039	2040	2041	2042	2043	2044	2045	2046	2047	2048	2049	2050	2051	2052	2053	2054	2055	2056	2057	2058	2059	2060	2061	2062	2063	2064	2065	2066	2067	2068	2069	2070	2071	2072	2073	2074	2075	2076	2077	2078	2079	2080	2081	2082	2083	2084	2085	2086	2087	2088	2089	2090	2091	2092	2093	2094	2095	2096	2097	2098	2099	2100	2101	2102	2103	2104	2105	2106	2107	2108	2109	2110	2111	2112	2113	2114	2115	2116	2117	2118	2119	2120	2121	2122	2123	2124	2125	2126	2127	2128	2129	2130	2131	2132	2133	2134	2135	2136	2137	2138	2139	2140	2141	2142	2143	2144	2145	2146	2147	2148	2149	2150	2151	2152	2153	2154	2155	2156	2157	2158	2159	2160	2161	2162	2163	2164	2165	2166	2167	2168	2169	2170	2171	2172	2173	2174	2175	2176	2177	2178	2179	2180	2181	2182	2183	2184	2185	2186	2187	2188	2189	2190	2191	2192	2193	2194	2195	2196	2197	2198	2199	2200	2201	2202	2203	2204	2205	2206	2207	2208	2209	2210	2211	2212	2213	2214	2215	2216	2217	2218	2219	2220	2221	2222	2223	2224	2225	2226	2227	2228	2229	2230	2231	2232	2233	2234	2235	2236	2237	2238	2239	2240	2241	2242	2243	2244	2245	2246	2247	2248	2249	2250	2251	2252	2253	2254	2255	2256	2257	2258	2259	2260	2261	2262	2263	2264	2265	2266	2267	2268	2269	2270	2271	2272	2273	2274	2275	2276	2277	2278	2279	2280	2281	2282	2283	2284	2285	2286	2287	2288	2289	2290	2291	2292	2293	2294	2295	2296	2297	2298	2299	2300	2301	2302	2303	2304	2305	2306	2307	2308	2309	2310	2311	2312	2313	2314	2315	2316	2317	2318	2319	2320	2321	2322	2323	2324	2325	2326	2327	2328	2329	2330	2331	2332	2333	2334	2335	2336	2337	2338	2339	2340	2341	2342	2343	2344	2345	2346	2347	2348	2349	2350	2351	2352	2353	2354	2355	2356	2357	2358	2359	2360	2361	2362	2363	2364	2365	2366	2367	2368	2369	2370	2371	2372	2373	2374	2375	2376	2377	2378	2379	2380	2381	2382	2383	2384	2385	2386	2387	2388	2389	2390	2391	2392	2393	2394	2395	2396	2397	2398	2399	2400	2401	2402	2403	2404	2405	2406	2407	2408	2409	2410	2411	2412	2413	2414	2415	2416	2417	2418	2419	2420	2421	2422	2423	2424	2425	2426	2427	2428	2429	2430	2431	2432	2
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5 It is another object of the present invention to provide a novel method and system for echo canceling, which alleviates the disadvantages of the prior art.

The signal processor determines an Euclidean norm of an
15 echo-replica signal. The signal processor also determines an
Euclidean norm of an input signal. The echo replica signal can be
produced by processing a reference far-end signal with the adaptive
filter. The adaptation control unit corrects the coefficients of the
adaptive filter. The adaptive filter derives an updated echo-replica
20 signal. The subtractor subtracts the updated echo-replica signal from
the input signal and derives an updated error signal thereof.

The echo canceling apparatus can further include a digital-to-analog converter, connected to the signal processor, to the subtractor and to a source of the input signal such as a microphone. The analog-to-digital converter converts an analog input signal into a digital input signal.

In accordance with another aspect of the present invention, there is thus provided a echo canceling method, including the steps of; determining an Euclidean norm of an echo-replica signal, determining an Euclidean norm of an input signal, determining a gradient step size, correcting the coefficients of an adaptive filter, deriving an updated echo-replica signal, and determining an updated error signal by subtracting the updated echo-replica signal from the input signal. The echo replica signal can be produced by processing a reference far-end signal with an adaptive filter.

The input signal can include at least one of the list consisting of an echo-signal, a near-end speech signal, a noise signal and the like. The gradient step size can be a function of the Euclidean norms of the echo-replica and the input signals. According to one aspect of the invention, the value of the function decreases when the values of the near-end speech and noise signals increase,

and the value of the function increases when the values of the near-end speech and noise signals decrease.

The correction of the coefficients of the adaptive filter can be performed by employing the gradient step size, the reference far-end signal and an error signal. According to a preferred embodiment
5 of the present invention, the error signal can be determined at the previous adaptation step.

In accordance with a further aspect of the present invention, there is thus provided an echo suppression apparatus
10 including a first amplitude estimation unit, a second amplitude estimation unit, a comparison unit, connected to the first amplitude estimation unit and to the second amplitude estimation unit, and a decision logic unit connected to the comparison unit.

The first amplitude estimation unit estimates an amplitude
15 measure of a reference far-end signal. The second amplitude estimation unit estimates an amplitude measure of an error signal. The comparison unit compares between the values of the amplitude measure of the far-end signal and the amplitude measure of the error signal and produces a comparison result thereof.

20 The decision logic unit analyzes the comparison result and produces a control signal thereof. The error signal can be an output

unit, connected to the comparison unit, and a decision logic unit connected to the delay unit.

The first amplitude estimation unit produces at least two far-end signal amplitude measure values by estimating an amplitude
5 measure of a reference far-end signal. The reference far-end signal is a sequence of at least two digital signal blocks, where each of the digital signal blocks contains at least one digital sample. The second amplitude estimation unit produces at least two error signal
10 amplitude measure values by estimating an amplitude measure of an error signal. The error signal is a sequence of at least two digital signal blocks, where each of the digital signal blocks contains at least one digital sample.

The comparison unit determines a plurality of value pairs. Each of the pairs includes a selected one of the error signal
15 amplitude measure values and a respective one of the far-end signal amplitude measure values. The comparison unit further compares between the far-end signal amplitude measure value and the error signal measure value, within each of the value pairs and produces at least two comparison results thereof.

The delay unit stores the comparison results. The decision logic unit analyzes the comparison results and produces a control signal thereof.

In accordance with yet another aspect of the present invention, there is provided an echo suppression method including the steps of:

- ξ Producing at least two far-end signal amplitude measure values by estimating an amplitude measure of a reference far-end signal. The reference far-end signal is a sequence of at least two digital signal blocks. Each of the digital signal blocks contains at least one digital sample.
- ξ Producing at least two error signal amplitude measure values by estimating an amplitude measure of an error signal. The error signal includes a sequence of at least two digital signal blocks. Each of the digital signal blocks contains at least one digital sample.
- ξ Determining a plurality of value pairs, where each pair includes a selected one of the error signal amplitude measure values and a respective one of the far-end signal amplitude measure values.

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ξ Comparing between the far-end signal amplitude measure value and the error signal amplitude measure value, within each of the value pairs, thereby producing at least two comparison results.

5 ξ Analyzing the at least two comparison results, thereby producing a control signal.

In accordance with another aspect of the present invention, there is thus provided an echo suppression apparatus which includes a first amplitude estimation unit, a second amplitude estimation unit, a comparison unit, connected to the first amplitude estimation unit and to the second amplitude estimation unit, and a decision logic unit, connected to the comparison unit.

10

The first amplitude estimation unit estimates an amplitude measure of a reference far-end signal. The second amplitude estimation unit estimates an amplitude measure of an error signal. The comparison unit compares between the values of the amplitude measure of the far-end signal and the amplitude measure of the error signal and produces a comparison result thereof.

15

The decision logic unit analyzes the comparison result and produces a control signal thereof. The error signal can be the output

20

amplitude measure values. The comparison unit compares between the far-end signal amplitude measure value and the error signal measure value, within each of the value pairs and produces at least two comparison results thereof. The delay unit stores the at least two comparison results. The decision logic unit analyzes the comparison results and produces a control signal thereof.

BRIEF DESCRIPTION OF THE DRAWINGS

The present invention will be understood and appreciated more fully from the following detailed description taken in conjunction with the drawings in which:

5 Figure 1 is a schematic illustration of a speech communication system, constructed and operative in accordance with a preferred embodiment of the present invention;

 Figure 2 is a schematic illustration in detail of the echo-cancellation system of Figure 1, constructed and operative in
10 accordance with a further preferred embodiment of the present invention;

 Figure 3 is a schematic illustration of a method for operating the echo-cancellation system of Figure 2, operative in accordance with a further preferred embodiment of the present
15 invention;

 Figure 4 is a schematic illustration in detail of the echo-suppression system of Figure 1, constructed and operative in accordance with a further preferred embodiment of the present invention; and

20 Figure 5 is a schematic illustration of a method for operating the echo-suppression system of Figure 4, operative in

accordance with another preferred embodiment of the present invention.

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DETAILED DESCRIPTION OF PREFERRED EMBODIMENTS

The present invention alleviates the disadvantages of the prior art by providing better robustness of echo cancellation, whenever an additive background noise and/or near-end speech are present. It also provides a more accurate and computation-effective method of echo suppression.

Reference is now made to Figure 1, which is a schematic illustration of a speech communication system, generally referenced 10, constructed and operative in accordance with a preferred embodiment of the present invention.

Speech communication system 10 includes a receiver 12, a loudspeaker 14, a switch 16, an echo-cancellation system 18, a transmitter 20, an echo-suppression system 22, a comfort noise generator 24, a digital-to-analog converter (D/A) 36 and a microphone 26.

Echo-cancellation system 18 is connected to receiver 12, to digital-to-analog converter 36, to echo-suppression system 22, and to microphone 26. Echo-suppression system 22 is connected to D/A 36, and to switch 16. Transmitter 20 and comfort noise generator 24 are connected to switch 16. Loudspeaker 14 is connected to D/A 36.

Receiver 12 receives an RF signal, extracts the audio information embedded therein and produces a digital audio stream X_k . Receiver 12 further provides the digital audio stream X_k to D/A 36, to echo-suppression system 22 and to echo-cancellation system 18. D/A 36 converts the digital audio stream X_k into an analog signal $x(t)$ and provides it to loudspeaker 14. Loudspeaker 14 converts further the signal $x(t)$ into an analog sound signal and provides it to a near-end speaker 30. Microphone 26 detects an echo-signal $y'(t)$, which is a combination of the sound, radiated by loudspeaker 14 and a sound, reflected from a reflector 32. Microphone 26 detects also a speech signal $v'(t)$ from the near-end speaker 30 and a noise signal $n'(t)$ from a noise source 28. Microphone 26 converts acoustic signals $y'(t)$, $v'(t)$, and $n'(t)$ into electric signals $y(t)$, $v(t)$ and $n(t)$, respectively. Microphone 26 provides a signal $d(t)$, which is a combination of the signals $y(t)$, $v(t)$ and $n(t)$ to echo-cancellation system 18.

Echo-cancellation system 18 analyzes the signals X_k and $d(t)$, estimates an echo replica and performs echo cancellation. The echo replica estimation is performed by employing a novel adaptive filtering method, which will be described in detail hereinafter. Echo-

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cancellation system 18 provides an output digital signal E_k to echo-suppression system 22.

Echo-suppression system 22 analyzes the signals from the loudspeaker 14 and echo-cancellation system 18. Based on
5 predetermined conditions, echo-suppression system 22 replaces, if necessary, the output signal E_k with an artificial noise signal 34, generated by comfort noise generator 24. Echo-suppression system 22 controls switch 16, which performs the switching between the output signal E_k and the artificial noise signal. Switch 16 provides the
10 output signal to transmitter 20.

Reference is now made to Figure 2, which is a schematic illustration in detail of echo-cancellation system 18 (Figure 1), constructed and operative in accordance with a further preferred embodiment of the present invention.

15 Echo-cancellation system 18 includes analog-to-digital converter (A/D) 58, an adaptive filter 54, an adaptation control unit 62, a signal processor 60, and a summator 56. Adaptive filter 54 is connected to the first input of summator 56, to signal processor 60, and to adaptation control unit 62. A/D converter 58 is connected to
20 the second input of summator 56, and to signal processor 60.

Adaptation control unit 62 is further connected to the output of summator 56, and to signal processor 60.

A/D converter 58 converts an analog signal $d(t)$ into a digital signal D_k . System 18 operates on digital signal blocks of length L , hence the signal D_k is a vector of length L , with components $D_k(0), D_k(1) \dots D_k(L-1)$. Index k denotes a block number and has values $0, 1, 2, \dots$. The signal $d(t)$ is a sum of signals $y(t)$, $v(t)$ and $n(t)$. Hence, signal D_k will be a sum of respective digital signals Y_k , V_k and N_k , each of them being a vector of length L . Accordingly, the signal X_k is a vector of length L , with components $X_k(0), X_k(1) \dots X_k(L-1)$.

Signal processor 60 receives near-end signal D_k and echo-replica signal \hat{Y}_k , determines Euclidean norms of the signals D_k and \hat{Y}_k , and provides it to adaptation control unit 62. It is noted that the filter tap values, which are used for echo-replica signal \hat{Y}_k calculation, were determined during the previous iteration.

Adaptation control unit 62 receives far-end signal X_k , error (residual) signal $E_k = D_k - \hat{Y}_k$ and the Euclidean norms of the signals D_k and \hat{Y}_k , and outputs filter tap correction values thereof.

Adaptation control unit 62 provides further corrected filter tap values to adaptive filter 54.

Adaptive filter 54 receives the far-end signal X_k and the corrected filter tap values, and outputs echo-replica signal \hat{Y}_k .

5 Adaptive filter 54 provides echo-replica signal \hat{Y}_k to summator 56.

Summator 56 sums signals D_k and \hat{Y}_k , produces an error signal E_k at the output, and provides signal E_k to echo-suppression system 22, and to adaptation control unit 62.

Reference is further made to Figure 3, which is a
10 schematic illustration of a method for operating echo-cancellation
system 18 (Figure 2), operative in accordance with a further
preferred embodiment of the present invention.

In step 70, a far-end signal X_k is received by adaptation control unit 62, and adaptive filter 54. With reference to Figure 2, a digital far-end signal X_k is provided by a far-end speaker to adaptation control unit 62, and to adaptive filter 54.

In step 72, a near-end signal D_k is received by summator 56, and signal processor 60. With reference to Figure 2, A/D converter 58 converts an analog near-end signal $d(t)$ into a digital near-end signal D_k , and provides it to summator 56, and to signal processor 60.

In step 74, Euclidean norms $\|D_k(n)\|$ and $\|Y_k(n)\|$ of the signals D_k and \hat{Y}_k respectively, at time instance n , are calculated. With reference to Figure 2, signal processor 60 receives signals D_k and \hat{Y}_k , and determines Euclidean norms $\|D_k(n)\|$ and $\|\hat{Y}_k(n)\|$

5 according to the following expressions:

$$\|D_k(n)\|^2 = \sum_{l=0}^{L-1} D_k^2(n-l), \quad \|\hat{Y}_k(n)\|^2 = \sum_{l=0}^{L-1} \hat{Y}_k^2(n-l),$$

(1)

Signal processor 60 provides further the Euclidean norms $\|D_k(n)\|$ and $\|Y_k(n)\|$ to adaptation control unit 62.

10 In step 76, a gradient step size $\Pi(n)$, at time instance n , is determined. With reference to Figure 2, adaptation control unit 62 receives far-end signal block X_k , residual signal block E_k , and the Euclidean norms $\|D_k(n)\|$ and $\|\hat{Y}_k(n)\|$. Gradient step size $\Pi(n)$ is further determined as:

$$15 \quad \Pi(n) = \frac{\Pi_0 \|X_k(n)\| \frac{\Delta}{E} \|\hat{Y}_k(n)\|^2}{\|X_k(n)\|^2 \frac{\Delta}{E} \|\hat{Y}_k(n)\|^2 + \|D_k(n)\|^2 \Delta \|\hat{Y}_k(n)\|^2}, \quad (2)$$

where Δ and E are constants. The value of Δ depends on hardware and software implementation of the system, and can be determined

experimentally. The optimal value of Δ will be the one which effects in a maximal echo cancellation. The value of E can also be determined experimentally. The constant Δ is dimensionless, and the constant E has dimensions of energy. It is noted, that constants Δ and E can be multi-valued, i.e. they can be represented with vectors, having components $\Delta(i)$, $E(j)$ ($i, j = 1, 2, 3, \dots$). The most appropriate values of $\Delta(i)$, $E(j)$ can be selected, depending on the working conditions, and HW and SW implementation. It is appreciated, that other expressions for the gradient step size can be used, such as that proposed in the article by A. Hirano, et al. "A noise-robust stochastic gradient algorithm with an adaptive step-size suitable for mobile hands-free telephones", Proc. ICASST-95, v.5, pp. 1392-1395, 1995.

It is noted, that the value of the gradient step size $\mu(n)$ is controlled dynamically. It follows from equation (2), that $\mu(n)$ decreases with a rise in the near-end signal V_k or the noise signal N_k . In other words, the adaptation process will be performed more accurately, and hence, the robustness of the system will improve. In an opposite case of low near-end and noise signals, $\mu(n)$ will rise, causing, in turn, a rise in adaptation speed.

Based on the gradient step size $\mu(n)$, adaptation control unit 62 further determines an updated vector $H(n+1)$, at a time $n+1$. The components $H_q(n+1)$ of the vector $H(n+1)$ represent filter tap coefficients and are determined according to the following recurrent expression:

$$H_q(n+1) = H_q(n) - \mu(n) E_k(n) X_k(n-q), \quad (3)$$

where $q = 0, 1, 2, \dots, Q-1$, and Q is the order of adaptive filter 54. It is noted that the length L of the signal block X_k can be less than the order Q of adaptive filter 54. In this case, additional $Q - L$ zero valued samples must be appended to the signal block X_k , so that the dimension of the signal block X_k will be equal to dimension of the adaptive filter 54. Adaptation control unit 62 provides the updated vector H to adaptive filter 54.

In step 80, adaptive filter tap coefficients are updated and an echo-replica signal \hat{Y}_k is derived thereof. With reference to Figure 2, adaptive filter 54 receives near-end signal block X_k and vector H , and derives the echo-replica signal $\hat{Y}_k(n)$, at a time instance n , according to the following expression:

$$Y_k(n) = \int_{q=0}^{Q-1} H_k(q) X_k(n-q),$$

Comparison unit 108 is connected to delay unit 110, and to amplitude estimation units 104 and 106. Decision logic unit 112 is connected to delay unit 110.

Amplitude estimation units 104 and 106 receive signals E_k and X_k , respectively. Each of the amplitude estimation units further determines an amplitude estimation $A_E^{(k)}$ and $A_X^{(k)}$ of the respective signals, and provide the result to comparison unit 108.

Comparison unit 108 performs the comparison of amplitude estimations $A_E^{(k)}$ and $A_X^{(k)}$, and provides the result to delay unit 110. Delay unit 110 already contains the results of comparisons made for the previous $M-1$ pairs of signal blocks. The output of delay unit 110 is a vector \mathfrak{A}_k with components $\mathfrak{A}_k(0), \mathfrak{A}_k(1), \dots, \mathfrak{A}_k(M-1)$. Delay unit 110 provides vector \mathfrak{A}_k to decision logic unit 112.

Decision logic unit 112 analyzes vector \mathfrak{A}_k , and produces a control signal P . Decision logic unit 112 provides the control signal P to switch 16 (Figure 1).

Reference is further made to Figure 5, which is a schematic illustration of a method for operating echo-suppression system 22 (Figure 4), operative in accordance with another preferred embodiment of the present invention.

In steps 150 and 152, the far-end signal X_k and error signal E_k are received. With reference to Figure 4, amplitude estimation unit 106 receives the far end signal X_k , and amplitude estimation unit 104 receives error signal E_k from echo-cancellation system 18 (Figure 1).

5 In step 154, amplitude measures of the far-end and residual signals are estimated. With reference to Figure 4, amplitude estimation units 104 and 106 determine amplitude measures $A_E^{(k)}$ and $A_X^{(k)}$ for signals E_k and X_k , respectively, according to the following expressions

$$10 \quad A_E^{(k)} = \sum_{l=0}^{L-1} |E_k(l)|, \quad A_X^{(k)} = \sum_{l=0}^{L-1} |X_k(l)|, \quad (5)$$

where L is a length of signal blocks E_k and X_k . Amplitude measures $A_E^{(k)}$ and $A_X^{(k)}$ are further provided to comparison unit 108.

It is appreciated that instead of amplitude measures $A_E^{(k)}$ and $A_X^{(k)}$, it is possible to use other signal measures, for example
 15 energy estimate. It is noted, that in distinction to energy estimation, which is quadratic in signal amplitude, amplitude measures $A_E^{(k)}$ and $A_X^{(k)}$ are linear functions of signal amplitudes E_k and X_k , respectively. The use of $A_E^{(k)}$ and $A_X^{(k)}$ reduces the amount of the required calculations, and furthermore, increases their accuracy. In step 156,
 20 amplitude measures $A_E^{(k)}$ and $A_X^{(k)}$ are compared. With reference to

Figure 4, comparison unit 108 receives amplitude measures $A_E^{(k)}$ and $A_X^{(k)}$, compares them, and produces a boolean output $\vartheta_k(0)$, according to the following conditional expressions:

$$\begin{aligned} \uparrow \vartheta_k(0) &= 1, \text{ if } A_X^{(k)} - A_E^{(k)} \geq T, \\ \rightarrow \vartheta_k(0) &= 0, \text{ otherwise} \end{aligned} \quad (6)$$

- 5 where T is a threshold value. The value of T is determined experimentally. The value of $\vartheta_k(0)$ is further provided to delay unit 110. Delay unit 110 already contains the results of comparisons made for the previous $M-1$ pairs of signal blocks. The output of delay unit 110 is a vector ϑ_k with components $\vartheta_k(0), \vartheta_k(1), \dots, \vartheta_k(M-1)$.
- 10 Delay unit 110 provides vector ϑ_k to decision logic unit 112.

- In step 158, the results of the current comparison, and previous $M-1$ comparisons of amplitude measures, are analyzed, and a respective control signal is derived thereof. With the reference to Figure 4, decision logic unit 112 receives comparison vector ϑ_k ,
- 15 analyzes comparison vector ϑ_k , and derives appropriate control signal P for switch 16. The decision process is accomplished in accordance with the following logical scheme:

$$\begin{aligned} &\text{if } (\vartheta_k(0) == 1, \vartheta_k(1) == 1, \dots, \vartheta_k(M-1) == 1) \\ &\quad \text{count} = M \end{aligned}$$

endif
count=count-1
if(count>0)
P==1
5 *else*
P==0
endif

In step 160, control signal P is provided to switch 16. With reference to Figure 4, decision logic unit 112 provides control signal
10 P to switch 16. In case $P==1$, switch 16 replaces signal E_k with comfort noise, otherwise signal E_k is provided to transmitter 20.

It will be appreciated by persons skilled in the art that the present invention is not limited to what has been particularly shown and described hereinabove. Rather the scope of the present
15 invention is defined only by the claims, which follow.

CLAIMS

1. An echo canceling method, the method comprising the steps of:
- determining an Euclidean norm of an echo-replica signal;
 - determining an Euclidean norm of an input signal;
 - 5 determining a gradient step size;
 - correcting the coefficients of an adaptive filter;
 - deriving an updated echo-replica signal, and
 - determining an updated error signal by subtracting said
 - updated echo-replica signal from said input signal.
- 10
2. The method according to claim 1, wherein said echo replica signal is produced by processing a reference far-end signal with an adaptive filter.
- 15
3. The method according to claim 1, wherein said input signal includes at least one of the list consisting of:
- an echo-signal;
 - a near-end speech signal, and
 - a noise signal.
- 20

4. The method according to claim 1, wherein said gradient step size is a function of said Euclidean norms of said echo-replica and input signals.
5. The method according to claim 4, wherein the value of said function decreases when the values of said near-end speech and noise signals increase, and wherein the value of said function increases when the values of said near-end speech and noise signals decrease.
6. The method according to claim 1, wherein the correction of said coefficients of said adaptive filter is performed by employing said gradient step size, said reference far-end signal and an error signal, said error signal determined at the previous adaptation step.
7. Echo canceling apparatus, comprising:
 - a signal processor;
 - an adaptation control unit, connected to said signal processor;
 - an adaptive filter, connected to said signal processor and to said adaptation control unit, and

a subtractor, connected to said signal processor, to said adaptation control unit and to said adaptive filter.

8. The echo canceling apparatus, according to claim 7, wherein
5 said signal processor determines an Euclidean norm of an echo-replica signal.

9. The echo canceling apparatus, according to claim 7, wherein
said signal processor determines an Euclidean norm of an input
signal.

10
10. The echo canceling apparatus, according to claim 8, wherein
said echo replica signal is produced by processing a reference
far-end signal with said adaptive filter.

15 11. The echo canceling apparatus, according to claim 9, wherein
said input signal includes at least one of the list consisting of:

an echo-signal;

a near-end speech signal, and

a noise signal.

20

12. The echo canceling apparatus, according to claim 7, wherein said adaptation control unit determines a gradient step size.

13. The echo canceling apparatus, according to claim 12, wherein said gradient step size is a function of said Euclidean norms of said echo-replica and input signals.

14. The echo canceling apparatus, according to claim 13, wherein the value of said function decreases when the values of said near-end speech and noise signals increase and wherein the value of said function increases when the value of said near-end speech and noise signals decrease.

15. The echo canceling apparatus, according to claim 7, wherein said adaptation control unit corrects the coefficients of said adaptive filter.

16. The echo canceling apparatus, according to claim 15, wherein said correction of said coefficients of said adaptive filter is performed by employing said gradient step size, said reference

far-end signal and an error signal, said error signal determined at the previous adaptation step.

17. The echo canceling apparatus, according to claim 7, wherein
5 said adaptive filter derives an updated echo-replica signal.

18. The echo canceling apparatus, according to claim 7, wherein
said subtractor subtracts said updated echo-replica signal from
said input signal and derives an updated error signal thereof.
10

19. The echo canceling apparatus, according to claim 7, further
comprising a digital-to-analog converter, connected to said
signal processor, to said subtractor and to a source of said
input signal.
15

20. The echo canceling apparatus, according to claim 19, wherein
said analog-to-digital converter converts an analog input signal
into a digital input signal.

20 21. Echo suppression method comprising the steps of:

estimating an amplitude measure of a reference far-end
signal;

estimating an amplitude measure of an error signal;

comparing between the values of said amplitude measures of
said far-end and said error signals, thereby producing a
comparison value; and

analyzing said comparison value, thereby producing a
control signal.

22. The echo suppression method, according to claim 21, wherein
said error signal is received from an echo cancellation system.

23. The echo suppression method, according to claim 21, wherein
said far-end signal and said error signal, are digital signal
blocks, each said digital signal block containing at least one
digital sample.

24. An echo suppression method comprising the steps of:
producing at least two far-end signal amplitude measure
values by estimating an amplitude measure of a reference far-
end signal, wherein said reference far-end signal is a sequence

a comparison unit, connected to said first amplitude estimation unit and to said second amplitude estimation unit; and

a decision logic unit connected to said comparison unit.

5

26. The echo suppression apparatus, according to claim 25, wherein said first amplitude estimation unit estimates an amplitude measure of a reference far-end signal.

10 27. The echo suppression apparatus, according to claim 25, wherein said second amplitude estimation unit estimates an amplitude measure of an error signal.

28. The echo suppression apparatus, according to claim 25, wherein said first amplitude estimation unit estimates an amplitude measure of a reference far-end signal,

15 wherein said second amplitude estimation unit estimates an amplitude measure of an error signal, and

wherein said comparison unit compares between the values of said amplitude measure of said far-end signal and

20

said amplitude measure of said error signal and produces a comparison result thereof.

29. The echo suppression apparatus, according to claim 28,
5 wherein said decision logic unit analyzes said comparison result and produces a control signal thereof.

30. The echo suppression apparatus, according to claim 27,
10 wherein said error signal is an output of an echo cancellation system.

31. The echo suppression apparatus, according to claim 25,
wherein said first amplitude estimation unit estimates an
15 amplitude measure of a reference far-end signal,
wherein said second amplitude estimation unit estimates an amplitude measure of an error signal, and
wherein said far-end signal and said error signal are digital
signal blocks, each said digital signal block contains at least
20 one digital sample.

32. Echo suppression apparatus, comprising:

a first amplitude estimation unit ;

a second amplitude estimation unit;

5 a comparison unit, connected to said first amplitude estimation unit and to said second amplitude estimation unit;

a delay unit, connected to said comparison unit; and

a decision logic unit, connected to said delay unit.

10 33. The apparatus for echo suppression, according to claim 32, wherein said first amplitude estimation unit produces at least two far-end signal amplitude measure values by estimating an amplitude measure of a reference far-end signal, wherein said reference far-end signal is a sequence of at least two digital
15 signal blocks, each said digital signal block contains at least one digital sample.

34. The apparatus for echo suppression, according to claim 32, wherein said second amplitude estimation unit produces at
20 least two error signal amplitude measure values by estimating an amplitude measure of an error signal, wherein said error

signal is a sequence of at least two digital signal blocks, each said digital signal block contains at least one digital sample.

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35. The apparatus for echo suppression, according to claim 32,
5 wherein said comparison unit determines a plurality of value pairs, each said pair including a selected one of said at least two error signal amplitude measure values and a respective one of said at least two far-end signal amplitude measure values, compares between said far-end signal amplitude
10 measure value and said error signal measure value, within each said value pairs and produces at least two comparison results thereof.

36. The apparatus for echo suppression, according to claim 35,
15 wherein said delay unit stores said at least two comparison results.

37. The apparatus for echo suppression, according to claim 36,
20 wherein said decision logic unit analyzes said at least two comparison results and produces a control signal thereof.

ABSTRACT

Echo canceling method including the steps of determining an Euclidean norm of an echo-replica signal, determining an Euclidean norm of an input signal, determining a gradient step size, 5 correcting the coefficients of an adaptive filter, deriving an updated echo-replica signal, and determining an updated error signal by subtracting the updated echo-replica signal from the input signal.

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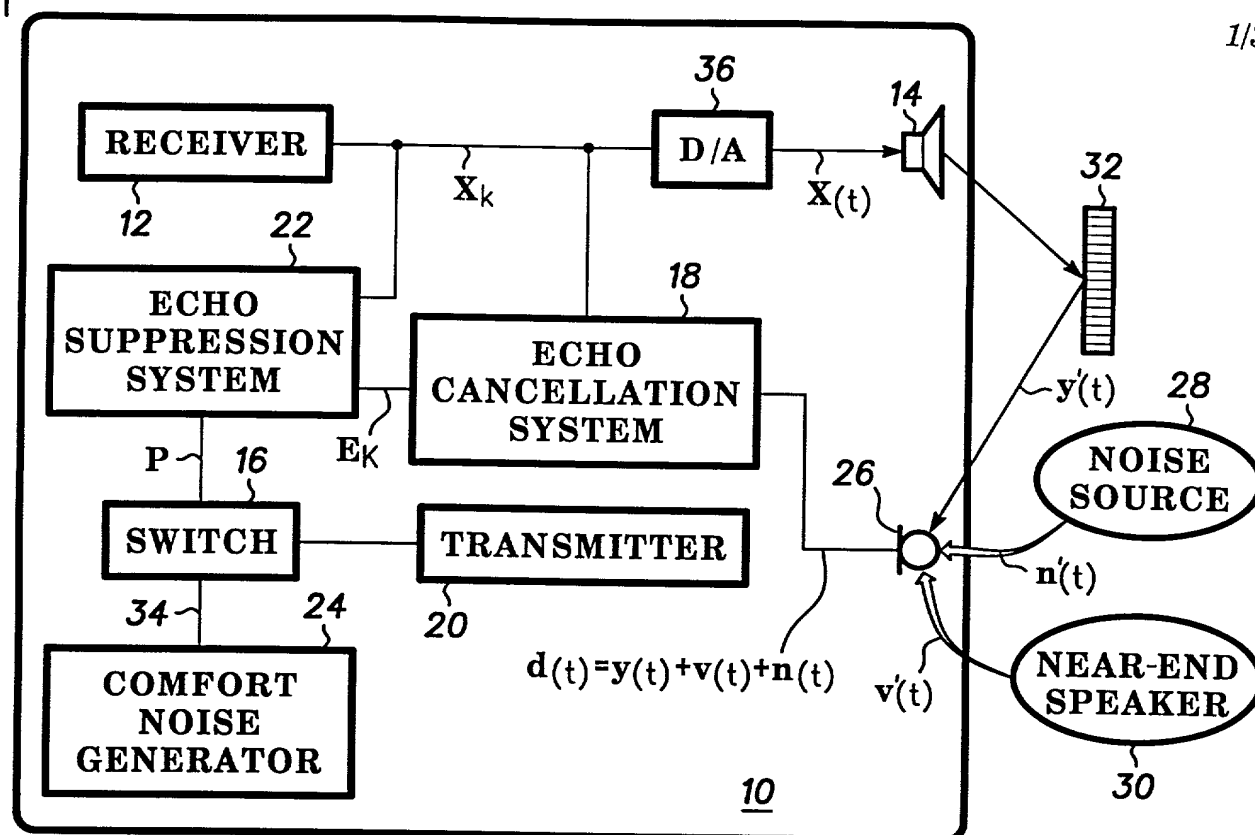


FIG. 1

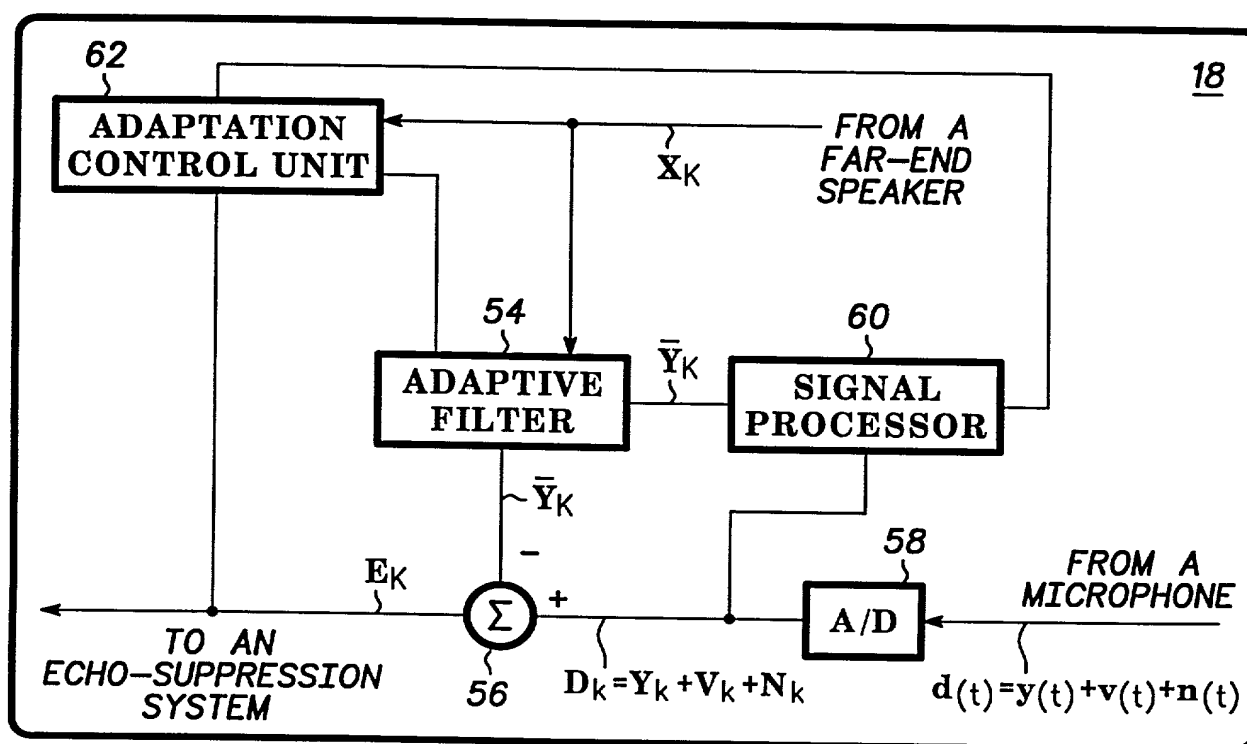
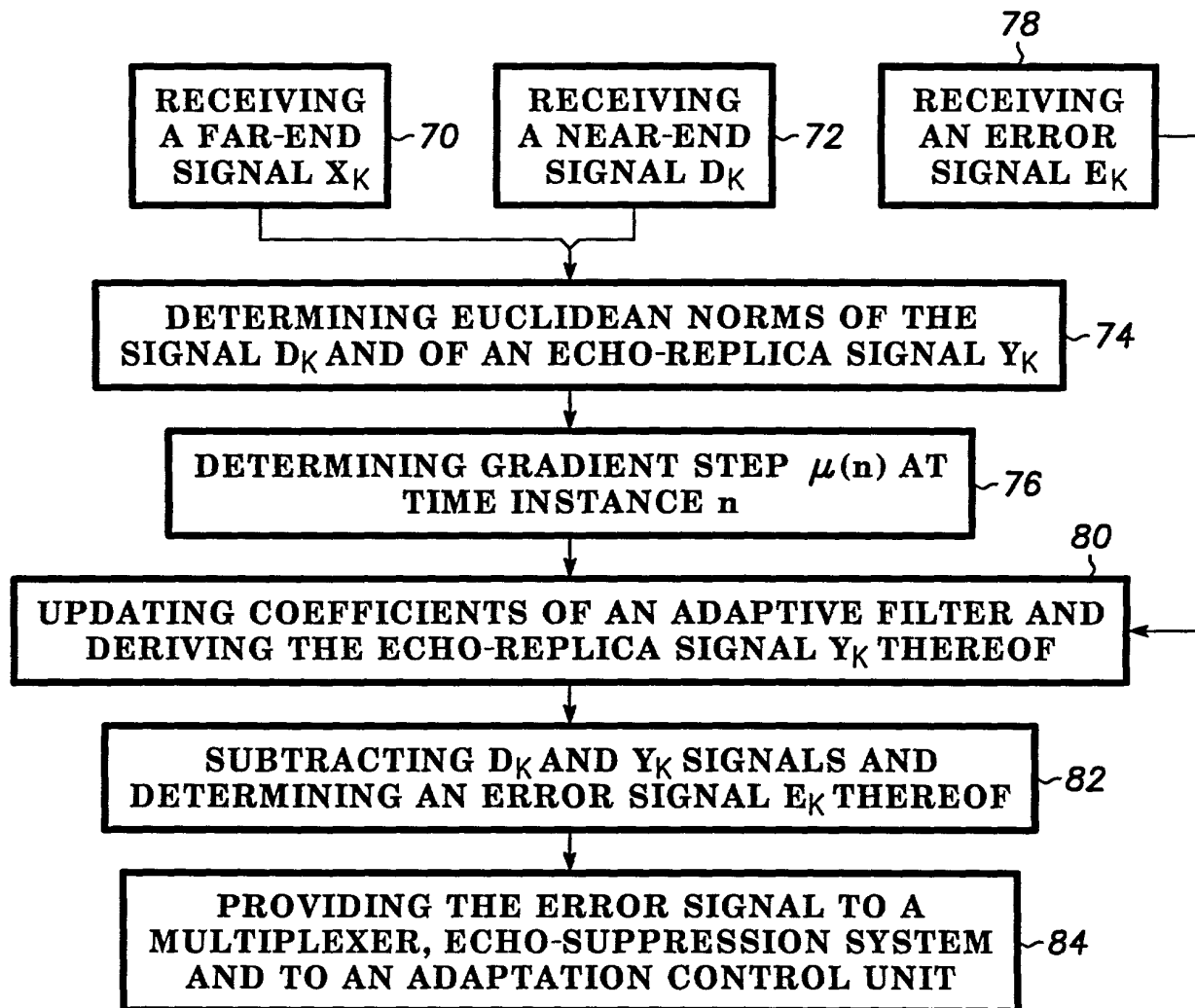


FIG. 2

*FIG. 3*

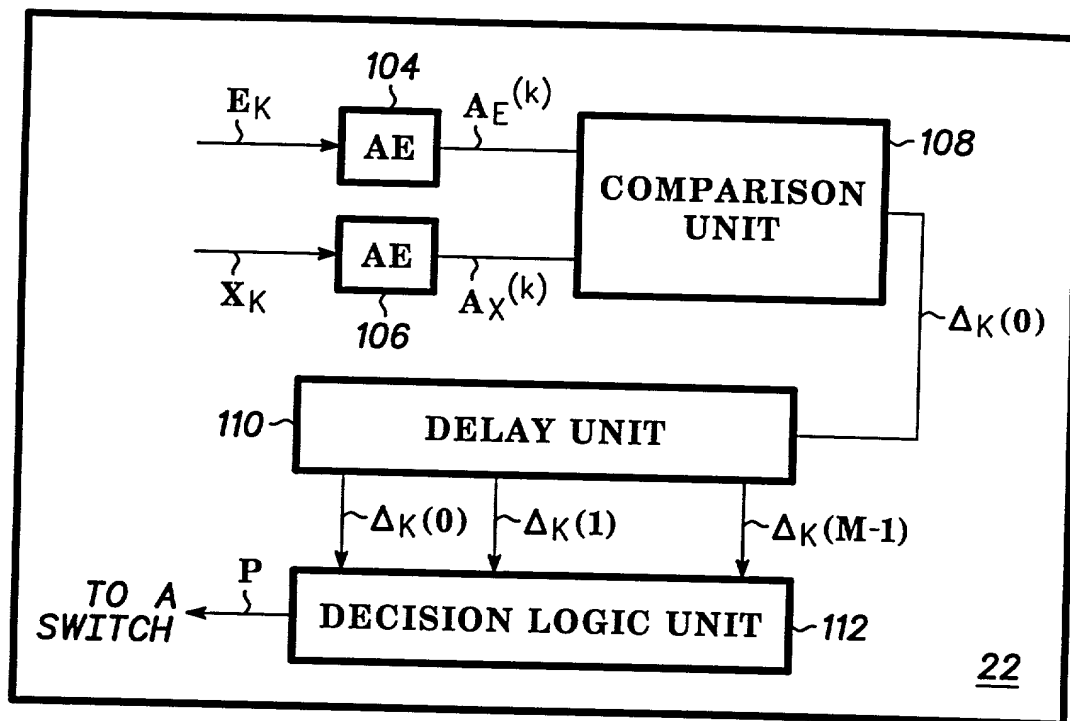


FIG. 4

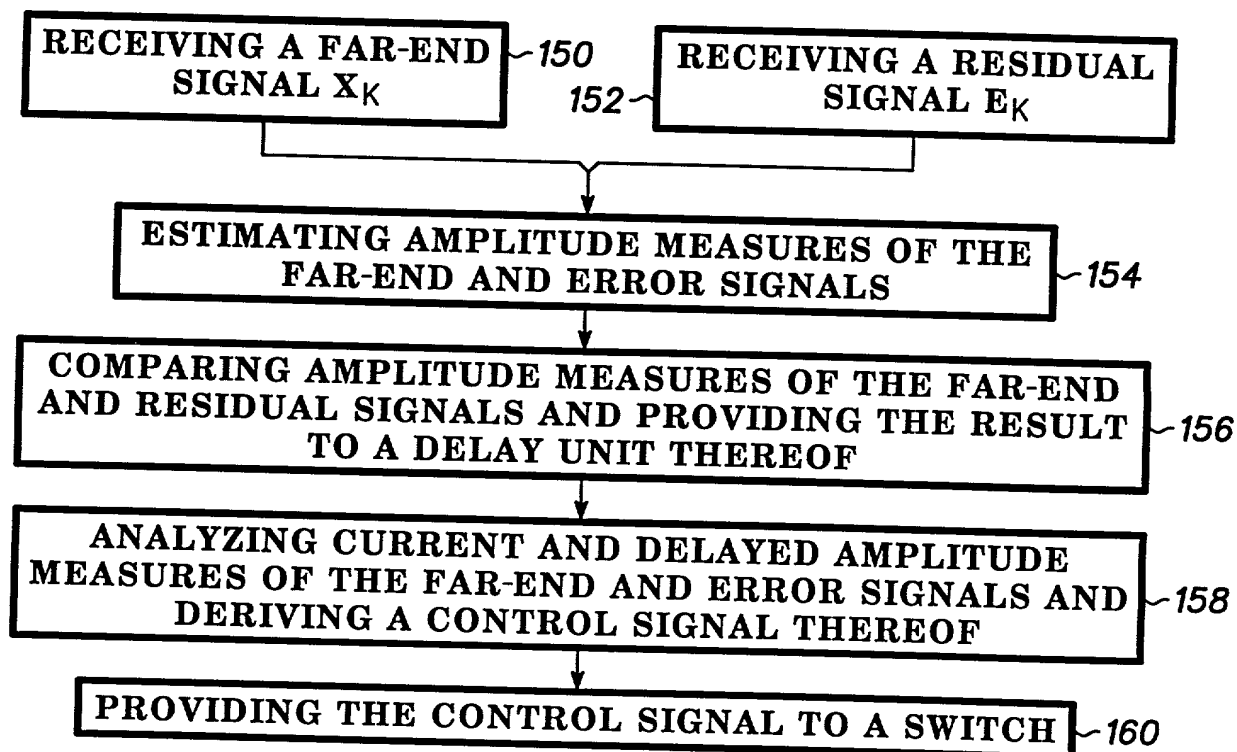


FIG. 5

[illegible]

As a below named inventor, I hereby declare that:

I believe I am the original, first and sole inventor (if only one name is listed below), or an original, first and joint inventor (if plural names are listed below), of the subject matter which is claimed and for which a patent is sought on the invention entitled ECHO SUPPRESSION AND ECHO CANCELLATION, the specification of which is attached hereto unless the following box is checked:

I hereby state that I have reviewed and understand the contents of the above identified specification, including the claims, as amended by any amendment referred to above.

I hereby claim foreign priority benefits under Title 35, United States Code, § 119(a)-(d) of any foreign application(s) for patent or inventor's certificate listed below and have also identified below any foreign application for patent or inventor's certificate having a filing date before that of the application on which priority is claimed.

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